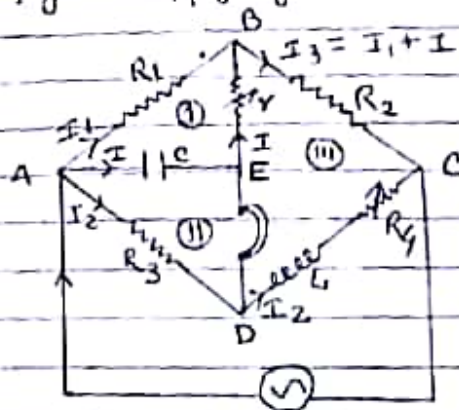


Anderson Bridge  
 A-2(11) Paper IV

This is the modified form of Maxwell's L/C bridge. The double balance is obtained by later adjusting resistances only, the standard capacitor being invariable. Anderson's bridge shown in fig is formed by placing the condenser  $C$  and a non inductive resistance  $r$  in series with the condenser, the combination being parallel with  $R_3$ ; the other components and detector are placed as shown in fig. At the time of balance  $E$  and  $D$  are at the same potential and let the current vectors in the different parts of the bridge be as indicated in fig. Applying Kirchhoff's law we have



$$R_1 I_1 - \left( r + \frac{1}{j\omega C} \right) I = 0 \quad [\text{for loop I}] \quad \text{--- (1)}$$

$$\frac{1}{j\omega C} \cdot I - R_3 I_2 = 0 \quad [\text{for loop II}] \quad \text{--- (2)}$$

$$R_2 (I_1 + I) + rI - (j\omega L + R_4) I_2 = 0 \quad [\text{for loop III}] \quad \text{--- (3)}$$

Substituting the value of  $I_1$  &  $I_2$  from equ<sup>n</sup> (1) & (2) in equ<sup>n</sup> (3) we have,

$$\frac{R_2}{R_1} \left( r + \frac{1}{j\omega C} \right) I + R_2 I - \frac{(j\omega L + R_4) I + rI}{j\omega C R_3} = 0 \quad \text{--- (4)}$$

Equating real & imaginary parts, we have

$$\frac{R_2}{R_1} r + R_2 - \frac{L}{CR_3} + r = 0$$

$$\text{or } L = CR_3 \left[ \frac{R_2}{R_1} r + R_2 + r \right]$$

$$= CR_3 \left[ r \left( 1 + \frac{R_2}{R_1} \right) + R_2 \right] \quad \text{--- (5)}$$

$$\text{and } \frac{R_2}{R_1 j\omega C} - \frac{R_4}{R_3 j\omega C} = 0 \text{ or } \frac{R_1}{R_2} = \frac{R_3}{R_4} \quad \text{--- (6)} \quad \underline{2}$$

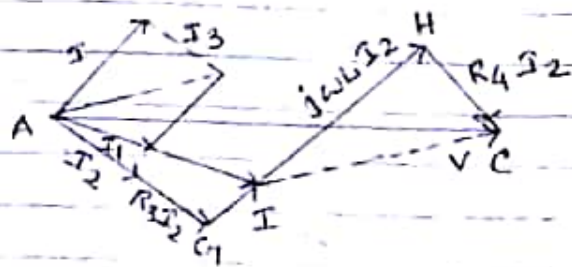
These conditions are independent of the frequency of the source and can be satisfied independently of each other by adjusting  $R_4$  in the latter case and  $r$  in the former case. Equn (5) shows that balance is possible only when  $L > CR_2R_3$  otherwise  $r$  will be negligible negative. Since there is no other inductance all mutual reactions between neighbouring inductance is avoided.

It is best to use  $R_1 : R_2 = 1 : 1$

The formula then reduces to

$$R_3 = R_4 \text{ and } L = CR_3(2r + R_2) \quad \text{--- (f)}$$

The vector diagram for the Anderson bridge is shown below. If  $AC$  is  $\vec{V}$ , the potential vector for the applied voltage, if  $I_2$  represents the current in  $ADC$ , then  $AC_1 = R_3 I_2$  and  $CH = j\omega L I_2$  and  $HC = R_4 I_2$ , so that  $HC \parallel$  to  $AC_1$  and



$CH \perp$  to  $AC_1$  or  $HC$ . Since the potential of  $E$  is the same as that of  $D$ , hence

$$V_E = V_{AD}$$

$$\text{or } \left(\frac{1}{j\omega C}\right) I = R_3 I_2$$

This relation shows that  $I$  is in quadrature with  $I_2$ .

If we draw  $O_1 I = R I$  then the vector  $A I$  is  $R_1 I_1$ .

The vector  $IC$  is  $R_2 I_3 = R_2 (I_1 + I)$ .

First of all the circuit is arranged as simple wheatstone Bridge connecting the inductance in the 4th arm  $ED$ . The resistance  $R_1$  and  $R_2$  are fixed to 10 ohm each and  $R_3$  is varied until the galvanometer deflection changes direction. The expected value is calculated with the formula (6), The expt. is

3.

repeated with  $\frac{R_1}{R_2} = 10$  and  $100$  and the actual resistance in the inductance arm  $R_2$  is calculated.

Now  $R_1$  and  $R_2$  are readjusted to the ratio  $10:10$  and a fractional resistance  $R'$  is placed in series with the inductance. Now  $R_3$  and  $R_4$ ,  $R'$  are varied for no deflection in the galvanometer. In this case

$R_2 + R' = R_4 = R_3$ . To increase the sensitivity the resistances are so arranged that  $R_1 = R_2 = \frac{1}{2} R_3$

Now galvanometer and D.C source are replaced by Headphone and A.C source respectively and capacitance  $C$  and resistance  $r$  are inserted as shown in fig.  $R_1, R_2, R_3$  and  $R_4$  are kept as such and the bridge is balanced for minimum sound with the help of resistance  $r$ . The expt. may be repeated by taking different capacitances. Substituting these values in equ<sup>n</sup> (7) self inductance of a given coil can be calculated.